

# **Random Wireless Network**

An Information Theoretic Perspective

**Rahul Vaze** 

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#### **CAMBRIDGE** UNIVERSITY PRESS

Cambridge House, 4381/4 Ansari Road, Daryaganj, Delhi 110002, India

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www.cambridge.org Information on this title: www.cambridge.org/9781107102323

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First published 2015

Printed in India

A catalogue record for this publication is available from the British Library

Library of Congress Cataloguing-in-Publication data Vaze, Rahul. Random wireless networks : an information theoretic perspective / Rahul Vaze. pages cm Includes bibliographical references and index. Summary: "Provides detailed discussion on single hop and multi hop model, feedback constraints and modern communication techniques such as multiple antenna nodes and cognitive radios"– Provided by publisher. ISBN 978-1-107-10232-3 (hardback) 1. Wireless communication systems. I. Title. TK5103.2.V39 2015 621.382'1–dc23 2014044738

ISBN 978-1-107-10232-3 Hardback

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To my son Niraad; this book was written while babysitting him.

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### Preface

In addition to the traditional cellular wireless networks, in recent past, many other wireless networks have gained widespread popularity, such as sensor networks, military networks, and vehicular networks. In a sensor network, a large number of sensors are deployed in a geographical area for monitoring physical parameters (temperature, rainfall), intrusion detection, animal census, etc., while in a military network, heterogenous military hardware interconnects to form a network in a battlefield, and vehicular networks are being deployed today for traffic management, emergency evacuations, and efficient routing. For efficient scalability, these new wireless networks are envisaged to be self-configurable with no centralized control, sometimes referred to as *ad hoc* networks.

The decentralized mode of operation makes it easier to deploy these networks, however, that also presents with several challenges, such as creating large amount of interference, large overheads for finding optimal routes, complicated protocols for cooperation and coordination. Because of these challenges, finding the performance limits, both in terms of the amount of information that can be carried across the network and ensuring connectivity in the wireless network, is a very hard problem and has remained unsolved in its full generality.

From an information-theoretic point of view, where we are interested in finding the maximum amount of information that can be carried across the network, one of the major bottlenecks in wireless network is the characterization of interference. To make use of the spatial separation between nodes of the wireless network, multiple transmitters communicate at the same time, creating interference at other receivers. The arbitrary topology of the network further compounds the problem by directly affecting the signal interaction or interference profile. Thus, one of the several trade-offs in wireless networks is the extent of spatial reuse viz-a-viz the interference tolerance. Another important trade-off is the relation between the radio range (distance to which each node can transmit) of sensor nodes and the connectivity of the wireless network. Small radio range leads to isolated nodes, while larger radio ranges result in significant interference at the neighboring receivers affecting connectivity.

Over the last decade and a half, these trade-offs have been addressed in a variety of ways, with exact answers derived for *random* wireless networks, where nodes of the wireless network are located uniformly at random in a given area of interest. The primary reason for assuming random location for nodes is the applicability of rich mathematical tools from stochastic geometry, percolation theory, etc. that provide significant mathematical foundation and allow derivation of concrete results. This book ties up the different ideas introduced for understanding the performance limits of random wireless networks and presents a complete overview on the advances made from an information-theoretic (capacity limits) point of view.

In this book, we focus on two capacity metrics for random wireless networks, namely, the transmission and the throughput capacity, that have been defined to capture the successful number

of bits that can be transported across the network. We present a comprehensive analysis of transmission capacity and throughput capacity of random wireless networks. In addition, using the tools from percolation theory, we also discuss the connectivity and percolation properties of random wireless networks, which impact the routing and large-scale connectivity in wireless networks. The book is presented in a cohesive and easy to follow manner, however, without losing the mathematical rigor. Sufficient background and critical details are provided for the advanced mathematical concepts required for solving these problems.

The book is targeted at graduate students looking for an easy and rigorous introduction to the area of information/communication theory of random wireless networks. The book also quantifies the effects of network layer protocols (e.g., automatic repeat requests (ARQs)), physical layer technologies such as multiple antennas (MIMO), successive interference cancelation, information-theoretic security, on the performance of wireless networks. The book is accessible to anyone with a background in basic calculus, probability theory, and matrix theory.

The book starts with an introduction to the signal processing, information theory, and communication theory fundamentals of a point-to-point wireless communication channel. Specifically, a quick overview of the concept of Shannon capacity, outage formulation, basic information-theoretic channels, basics of multiple antenna communication, etc. is provided that lays sufficient background for the rest of the book.

The book is divided into two parts, the first part exclusively deals with single-hop wireless networks, where each source-destination communicates directly with each other, while in the second part, we focus only on the multi-hop wireless networks, where source-destination pairs are out of each others' communication range and use multiple other nodes (called relays) for communication.

For the first part, we begin by deriving analytical expressions for the transmission capacity for a single-hop model with various scheduling protocols such as ALOHA, CSMA, guard-zone based, etc. Next, we discuss in detail the effect of using multiple antennas on the transmission capacity of a random wireless network and derive the optimal role of multiple antennas. We then extend our setup and present performance analysis of random wireless networks under a two-way communication model that allows for bidirectional communication between two nodes. We close the first part of the book by applying stochastic geometry tools to derive a tractable performance analysis of a cellular wireless network in terms of critical measures such as connection probability, average rate, etc. which is extremely useful for practicing engineers.

The second part of the book starts by extending the transmission capacity framework to a multihop wireless network, where we derive the transmission capacity expression and find the optimal value of several key parameters relevant to the multi-hop communication model. Then, we give a brief introduction to percolation theory results for both the discrete and the continuum case. The background on percolation theory sets up the platform for deriving several important results for random wireless networks, such as finding the optimal radio range for connectivity, formation of large connected clusters under different connection models, and most importantly for finding tight scaling bounds on the throughput capacity.

We then present the seminal result of Gupta and Kumar which shows that the throughput capacity of a random wireless network scales as square root of the number of nodes. Finally, we discuss the concept of hierarchical cooperation in a wireless network which is used to show that the throughput capacity can scale linearly with the number of nodes.

This book is an effort to present the several disparate ideas developed for deriving capacity of a random wireless network in a unified framework. For effective understanding, extensive effort is made to explain the physical interpretation of all results. As an attempt to reach out to a wider

audience, effects of practical communication models, such as cellular networks, two-way communication (downlink/uplink) feedback constraints, modern communication techniques (such as multiple antenna nodes, interference cancelation and avoidance, cognitive radios), are also analyzed and discussed in sufficient detail.

Most of the ideas/results presented in this book are not more than a decade old and have not yet found a consolidated treatment. The presentation is kept short and lucid with sufficient detail and rigor. For clarity, at instances, places simplified proofs of the original results are provided.

### Acknowledgments

I would like to thank everyone who have made this book possible. Starting with Srikanth Iyer, D. Yogeshwaran, Aditya Gopalan, Chandra Murthy, Radhakrishna Ganti, Sibiraj Pillai, Siddharth Banerjee, all have made detailed comments on my various drafts, which undoubtedly made the book more readable. Their critical comments have also shaped the structure and content. I would also like to thank Kaibin Huang, who urged me to write this as a textbook that he could use in his course. I do not know yet whether it will serve his purpose. Help from undergraduate internship students Vivek Bagaria, Dheeraj Narasimha, Jainam Doshi, Rushil Nagda, Siddharth Satpathi, Ajay Krishnan, and Navya Prem in proofreading is gratefully acknowledged. Their comments were valuable in making the book accessible to readers with little or no prior background. The review comments by the two referees helped immensely in changing the structure of the book. Their comments especially helped in the reorganization and pruning of the book to keep a clear and sharp focus.

### Notation

Α	Matrix A
$\mathbf{A}(i,j)$	(i, j)th entry of matrix A
a	vector a
$\mathbf{a}(i)$	<i>i</i> th element of vector <b>a</b>
$\mathbf{a}^{\dagger}$	Conjugate transpose of vector a
R	Set of real numbers
C	Set of complex numbers
$\mathbb{R}^d$	Set of real numbers in d dimensions
$\mathbb{C}^{m \times n}$	Matrices with $m$ rows and $n$ columns over the set of complex numbers
$\mathbb{P}(A)$	Probability of event A
E	Expectation operator
#(A)	Number of elements in set A
$\nu(A)$	Lebesgue measure of set A
0	Origin in $\mathbb{R}^d$
$d_{xy}$	Distance between nodes $x$ and $y$
α	Path-loss exponent for wireless propagation
$\mathbf{B}(x,r)$	Disc of radius $r$ centered at $x$
$\phi$	Null set
a	Absolute value of a
$\Phi$	A Poisson point process
p	ALOHA access probability
$\lambda$	Density of nodes of the network
$\mu$	Density of blockages/obstacles in the network
$\epsilon$	Outage probability constraint
Λ	Density measure of nodes of the network
$\gamma$	Interference suppression parameter
ρ	Random variable representing the random radius in the Gilbert's random disc
	model
$\Gamma(t)$	$\int_0^\infty x^{t-1} \exp^{-x} dx$
J	$\sqrt{-1}$
$A \propto B$	A = cB, where c is a constant
SNR	Signal-to-noise-ratio
SIR	Signal-to-interference-ratio
SINR	Signal-to-interference-plus-noise-ratio

β	Signal-to-interference-plus-ratio threshold for successful packet reception
В	Rate of transmission corresponding to SINR threshold $\beta$ , $\beta = 2^B - 1$
$M_n$	Number of retransmissions required on hop n
$N_h$	Number of hops
M	Number of end-to-end retransmissions required $\sum_{n=1}^{N_h} M_n = M$
t(n)	Per-node throughput capacity
T(n)	Network wide throughput capacity
C	Transmission capacity
$C_{\rm tw}$	Two-way transmission capacity
$C_d$	Delay normalized transmission capacity
$C_s$	Spatial progress capacity
AWGN	Additive white Gaussian noise
N <sub>0</sub>	Variance of the AWGN
$\mathcal{N}(m,var)$	Normal distribution with mean m and variance var
$\mathcal{CN}(m,var)$	Complex normal distribution with mean m and variance var
$1_n$	Indicator variable for node n
$\chi^2(2m)$	Chi-square distribution with $m$ degrees of freedom
$X \sim Y$	Random variable $X$ has distribution $Y$
f(n) =	If $\exists k > 0, n_0, \forall n > n_0,  g(n) k \le  f(n) $
$\Omega(g(n))$	
f(n) =	If $\exists k > 0, n_0, \forall n > n_0,  f(n)  \le  g(n) k$
$\mathcal{O}(g(n))$	
f(n) =	If $\exists k_1, k_2 > 0, n_0, \forall n > n_0,  g(n) k_1 \le  f(n)  \le  g(n) k_2$
$\Theta(g(n))$	
f(n) =	If $\lim_{n \to \infty} \frac{f(n)}{a(n)} = 0$
o(g(n))	<i>a</i> (

## Chapter 1 Introduction

#### **1.1 Introduction**

Wireless networks can be broadly classified into two categories: centralized and de-centralized. A canonical example of a centralized network is a cellular network, where all operations are controlled by basestations, for example, when should each user transmit or receive, thereby avoiding simultaneous transmission (interference) by closely located nodes. Prominent examples of de-centralized or *ad hoc* networks include sensor or military networks. Sensor network is deployed in a large physical area to either monitor physical parameters, such as temperature, rainfall, and animal census, or intrusion detection. In a military network, a large number of disparate military equipment, e.g., battle tanks, helicopters, ground forces, is connected in a decentralized manner to form a robust and high throughput network. Ad hoc networks are attractive because of their scalability, self-configurability, robustness, etc.

Vehicular network is a more modern example of an ad hoc wireless network, where a large number of sensors are deployed on the highways as well as mounted on vehicles that are used for traffic management, congestion control, and quick accident information exchange. Many other applications of ad hoc wireless networks are also envisaged such as deploying large number of sensors in large building for helping fire fighters in case of fire emergency and in case of earthquakes.

The key feature that distinguishes centralized and ad hoc wireless networks is interference. With centralized control, interference can be avoided in contrast to ad hoc networks, where there is no mechanism of inhibiting multiple transmitters from being active simultaneously. Thus, ad hoc networks give rise to complicated signal interaction at all receiver nodes. As compared to additive noise, interference is structured, and treating interference as noise is known to be sub-optimal. Thus, performance analysis of ad hoc wireless networks is far more complicated than centralized wireless networks.

In this book, we are interested in studying the physical layer issues of ad hoc wireless networks, such as finding the limits on the reliable rate of information transfer and ensuring connectivity among all nodes of the network. Traditionally, the Shannon capacity has been used to characterize the reliable rate of information transfer in communication systems. In a wireless network, however, finding the Shannon capacity is challenging and has remained unsolved. The main impediment in finding the Shannon capacity of wireless networks is the complicated nature of interference created by multiple simultaneously active transmitters at each other's receivers and network topology that directly influences the signal interaction.

To get some meaningful insights to the fundamental limits of throughput in wireless networks, alternate notions of capacity have been introduced and analyzed, such as transmission [1] and throughput/transport [2] capacity, which are defined by relaxing the reliability constraints compared to the Shannon capacity.

One key relaxation/assumption made for the purposes of analyzing these new capacity metrics is that the nodes of the network are assumed to be distributed uniformly at random in the area of interest, called the *random* wireless networks. The random node location assumption allows the use of tools from stochastic geometry and percolation theory for theoretical capacity analysis. In Chapter 2, we argue that random node location assumption is not very limiting for a practical ad hoc network.

Major focus of this book is on finding the transmission and the throughput capacity of random wireless networks. Through the transmission capacity formulation, we also quantify the effects of using multiple antennas at each node, using two-way communication between source and destination, effect of ARQ protocol, and using "smart" scheduling protocols in the random wireless networks. From here on in this book, when we say wireless network, we mean a random wireless network unless specified differently.

A necessary condition for finding the maximum rate of transmission or throughput between a pair of nodes in a wireless network is to ensure that they are connected to each other or have a connected path between each other, under a suitable definition of connection. Since any source can have an arbitrary choice for its destination, essentially, we need network wide connectivity, that is, each node pair should be reachable from every other node via connected paths. This condition is simply called as *connectivity* of the wireless network. Connectivity in a wireless network depends on the density of nodes, radio (transmission) range of any node, topology of the network, connection model between nodes, etc. In this book, we present relevant results from the percolation theory and then describe their application in finding the network parameters that ensure connectivity in wireless networks. Using percolation theory, we also study the size of the largest connected component in wireless networks and find conditions when the size of the largest connected component is a non vanishing fraction of the total number of nodes, which implies approximate connectivity.

The book is divided into two parts, first part exclusively deals with a single-hop model for wireless networks, where each source has a destination at a fixed distance from it and transmits its information directly to its destination without the help of any other node in the network. We define the notion of transmission capacity for the single-hop model and derive it for single antenna nodes, multiple antenna nodes, with scheduling protocols, and under two-way communication scenarios. The first part of the book also includes the performance analysis of cellular wireless network techniques using tools from stochastic geometry that are developed in the earlier chapters of the first part.

In the second part, we deal with the more relevant model of multi-hop communication for a wireless network and define two notions of capacity, namely the delay normalized transmission capacity and the throughput capacity and present their analysis. In addition, in the second part, we also study the connectivity and percolation properties of a multi-hop wireless network under the signal-to-noise-plus-interfence ratio (SINR) model.

This chapter sets up the background for studying wireless networks from a physical layer point of view. We begin by describing the basics of point-to-point communication, where a single transmitter is interested in communicating with a single receiver. To keep the discussion general, we consider the case when each node is equipped with multiple antennas. We first discuss the role of multiple antennas in improving the error-probability performance as a function of number of transmit and receive antennas with the optimal maximum likelihood (ML) decoder. We then state some difficulties in using the optimal maximum likelihood decoder, such as an exponential complexity, and present the more popular sub-optimal decoders such as zero-forcing decoder that have linear decoding complexity. We also discuss the error rate performance degradation while using the sub-optimal zero-forcing decoder.

Next, we define the notion of Shannon capacity and present results on the Shannon capacity of the point-to-point communication channel with multi-antenna equipped nodes. We show that the Shannon capacity scales linearly with the minimum of the number of transmit and receive antennas. We next present the outage formulation for characterizing capacity (called outage capacity) in non-ergodic channels, for which the Shannon capacity is zero. The non-ergodic channel is of interest since the popular slow-fading channel model of wireless signal propagation, where channel coefficients remain constant for sufficient amount of time, falls in the class of non-ergodic channels. The outage formulation also helps in defining the transmission capacity of wireless networks.

Next, we describe the received signal model at any node of a wireless network, where multiple transmitters are active at the same time. Using examples of some basic building blocks of a wireless network, we discuss some of the difficulties in finding the Shannon capacity of a wireless network. We then motivate the definitions of alternate capacity metrics, such as transmission capacity and throughput capacity, which are defined under a relaxed reliability constraint compared to the Shannon capacity.

We end this chapter by presenting some details on studying connectivity in wireless networks under various link connection models.

#### **1.2** Point-to-Point Wireless Signal Propagation Model

Consider a wireless communication channel between a single transmitter  $T_0$  equipped with  $N_t$  antennas and a single receiver  $R_0$  with  $N_r$  antennas. Let the distance between  $T_0$  and  $R_0$  be d, then the received signal at  $R_0$  at time t is given by

$$\mathbf{y}[t] = d^{-\alpha/2} \sqrt{\frac{P}{N_t}} \sum_{m=0}^{M-1} \mathbf{H}_m \mathbf{x}[t-m] + \mathbf{w}[t],$$
(1.1)

where M is the number of distinct multiple fading paths between the transmitter and the receiver,  $d^{-\alpha/2}$  is the distance-based path-loss function,  $\alpha$  is the path-loss exponent that is typically in the range (2, 4),  $\mathbf{H}_t \in \mathbb{C}^{N_r \times N_t}$  is the channel coefficient matrix at time t between the transmitter and the receiver, where  $\mathbf{H}_t(i, j)$  is the channel coefficient between the *i*th receive and *j*th transmit antenna. The  $N_t \times 1$  transmit signal vector at time t is  $\mathbf{x}[t]$  with unit power constraint,  $\mathbb{E}{\mathbf{x}[t]^{\dagger}\mathbf{x}[t]} = 1$ , Pis the average transmitted power, and  $\mathbf{w}[t]$  is additive white Gaussian noise vector with entries that are independent and  $\mathcal{CN}(0, 1)$  distributed.

**Assumption 1.2.1** Throughout this book, we will use the simple distance-based path-loss function of  $d^{-\alpha/2}$  that is valid in far-field, however, has a singularity in the near-field at d = 0.

**Assumption 1.2.2** We will also always assume a flat fading channel, that is, no multi-path  $\mathbf{H}_t = \mathbf{0}$  for t > 0, for which the signal model (1.1) simplifies to

$$\mathbf{y} = d^{-\alpha/2} \sqrt{\frac{P}{N_t}} \mathbf{H} \mathbf{x} + \mathbf{w}, \tag{1.2}$$

where the entries of **H** are assumed to be independent and  $\mathcal{CN}(0,1)$  distributed to model a rich scattering channel (Rayleigh fading). We also assume throughout this book that matrix **H** is perfectly known at the receiver.

To decode the transmit signal x, the optimal decoder is the maximum a posteriori (MAP) decoder that declares that signal to be transmitted which is the most likely signal x given the knowledge of y. Assuming an uniform distribution over the input signals, MAP decoding is equivalent to ML decoding, where the decoded codeword maximizes the likelihood of y given x. Mathematically, ML decoding solves the following optimization problem.

$$\max_{\mathbf{x}} \mathbb{P}(\mathbf{y}|\mathbf{x}, \mathbf{H}).$$

For the signal model (1.2), since each entry of w is independent and  $\mathcal{CN}(0,1)$  distributed,

$$\mathbb{P}(\mathbf{y}|\mathbf{x},\mathbf{H}) = \frac{1}{\pi} \exp^{-\left(\mathbf{y} - d^{-\alpha/2} \sqrt{\frac{P}{N_t}} \mathbf{H} \mathbf{x}\right) \left(\mathbf{y} - d^{-\alpha/2} \sqrt{\frac{P}{N_t}} \mathbf{H} \mathbf{x}\right)^{\dagger}},$$

which can be simplified to conclude that the ML decoder decodes vector x that solves

$$\max_{\mathbf{x}} \mathbb{P}(\mathbf{y}|\mathbf{x}, \mathbf{H}) = \min_{\mathbf{x}} ||\mathbf{y} - d^{-\alpha/2} \sqrt{\frac{P}{N_t}} \mathbf{H} \mathbf{x}||^2.$$
(1.3)

Thus, the ML decoder decodes  $\mathbf{x}$ , which is the closest codeword to the received signal  $\mathbf{y}$  in terms of the Euclidean distance. With ML decoding, all the components of vector  $\mathbf{x}$  are decoded jointly, thereby making the complexity exponential in the size of  $\mathbf{x}$  which is  $N_t$ .

Assuming that the channel matrix **H** remains constant for  $T \ge N_t$  time slots, and if the transmitter codes across T time slots to send codeword  $\mathbf{X}_i = [\mathbf{x}_i[1] \dots \mathbf{x}_i[T]]$ , the probability of decoding the codeword matrix  $\mathbf{X}_j = [\mathbf{x}_j[1] \dots \mathbf{x}_j[T]]$  instead of  $\mathbf{X}_i$  with an ML decoder is [3]

$$\mathbb{P}(\mathbf{X}_{i} \to \mathbf{X}_{j}) \leq \left(\prod_{k=1}^{\mathsf{div}} \sigma_{k} \left(\mathbf{X}_{i} - \mathbf{X}_{j}\right)\right)^{-N_{r}} P^{-\mathsf{div}N_{r}},\tag{1.4}$$

where

$$\operatorname{div} = \min_{\mathbf{X}_i \neq \mathbf{X}_j} \{ \operatorname{rank}(\mathbf{X}_i - \mathbf{X}_j) (\mathbf{X}_i - \mathbf{X}_j)^{\dagger} \}$$
(1.5)

and  $\sigma_k (\mathbf{X}_i - \mathbf{X}_j)$  are the non-zero eigenvalues of  $(\mathbf{X}_i - \mathbf{X}_j)(\mathbf{X}_i - \mathbf{X}_j)^{\dagger}$ . Thus, to minimize the pairwise error probability (1.4), one has to maximize the minimum of the rank of the difference of any two codeword matrices  $\mathbf{X}_i$  and  $\mathbf{X}_j$  (1.5). Clearly, with  $T \ge N_t$ , the maximum value of div is  $N_t$  (since  $\mathbf{X}_i \in \mathbb{C}^{N_t \times T}$ ,  $\forall i$ ) and for achieving div  $= N_t$ , the codewords  $\mathbf{X}_i$ 's should be coded in space and time; hence the codebook consisting of codewords  $\mathbf{X}_i$ 's is called a space-time block code (STBC). STBCs with div  $= N_t$  are called full-diversity achieving STBCs, and their error probability is proportional to  $P^{-N_tN_r}$ . Thus, with multiple transmit and receive antennas, the reliability of a wireless channel can be improved exponentially with the increasing transmission power.

Even though ML decoding provides with the best error probability performance, its decoding complexity is very high because of the joint decoding of all elements of transmitted vector  $\mathbf{x}$ . Several simple decoders with reduced decoding complexity are also known in literature, for example, minimum mean square error (MMSE) decoder and zero forcing (ZF) decoder. ZF decoder is specially attractive for its simple decoding rule and incurs linear decoding complexity in  $N_t$  (the number of elements of  $\mathbf{x}$ ). We describe the ZF decoder in brief and present its error probability performance. We will use the ZF decoder in Chapter 3 to analyze the effects of using multiple antennas in a wireless network.

With ZF decoder, to decode stream  $\mathbf{x}(\ell)$  of the transmitted vector

$$\mathbf{x} = [\mathbf{x}(1), \dots, \mathbf{x}(N_t)]^T,$$

the received signal (1.2) is multiplied with a vector  $\mathbf{q}_{\ell}^{\dagger} \in \mathbb{C}^{N_r \times 1}$ , which belongs to the null space of the columns  $\mathbf{H}(j), j = 1, \ldots, \ell - 1, \ell + 1, \ldots, N_t$  of the channel matrix  $\mathbf{H}$ , to cancel the inter-stream interference from all other streams  $\mathbf{x}(j), j = 1, \ldots, \ell - 1, \ell + 1, \ldots, N_t$ . With this operation, from (1.2), the resulting signal can be written as

$$\mathbf{y}(\ell) = d^{-\alpha/2} \sqrt{\frac{P}{N_t}} \mathbf{q}_{\ell}^{\dagger} \mathbf{H}(\ell) \mathbf{x}(\ell) + \mathbf{q}_{\ell}^{\dagger} \mathbf{w}, \qquad (1.6)$$

 $\forall \ell = 1, \dots, N_t$ , where there is no inter-stream interference from  $\mathbf{x}(j), j = 1, \dots, \ell - 1, \ell + 1, \dots, N_t$ . Thus, with a ZF decoder, each of the  $N_t$  data streams of  $\mathbf{x}$  can be decoded independently of each other using (1.6), thereby incurring linear decoding complexity compared to the exponential decoding complexity of the ML decoder. This sub-optimal receiver, however, has poor error probability performance because of correlating the noise components in  $y_\ell$  for different  $\ell = 1, \dots, N_t$ , and the error probability is proportional to  $P^{N_r - N_t + 1}$  [4], instead of  $P^{-N_t N_r}$  with the ML decoder.

We next discuss the alternative use of multiple antennas in improving the capacity of the pointto-point communication channel. We first define the concept of Shannon capacity, a measure of reliable throughput and show that Shannon capacity increases linearly with the minimum of the number of transmit and receive antennas.

#### **1.3 Shannon Capacity**

**Definition 1.3.1** The Shannon capacity C for a communication channel is defined as the largest quantity such that for any rate R < C, reliable communication is possible. By reliable communication, we mean that the probability of error can be driven down to zero with increasing block length. Conversely, if the rate of transmission  $R \ge C$ , the probability of error is lower bounded by a constant.

**Definition 1.3.2** Let x[n] and y[n] be the input and output of a channel at time n, respectively, then a channel is called a discrete memoryless channel (DMC), if given the most recent input, the output is independent of all previous inputs and outputs, that is,

 $\mathbb{P}(y[n] \mid x[1], \dots, x[n], y[1], \dots, y[n-1]) = \mathbb{P}(y[n] \mid x[n])$ 

for n = 1, 2, 3, ... Thus, in a DMC, given the input at time n, the output at time n is independent of all the past inputs and outputs.

C. E. Shannon, in his 1948 seminal paper [5], proved that the capacity of a DMC defined by  $\mathbb{P}(\mathbf{y}|\mathbf{x})$ , with input  $\mathbf{x} = [x[1], \dots, x[n]]$  and output  $\mathbf{y} = [y[1], \dots, y[n]]$  is given by

$$C = \max_{\mathbb{P}(\mathbf{x})} \mathsf{I}(\mathbf{x}; \mathbf{y}), \tag{1.7}$$

where l(x; y) is the mutual information between x and y [17]. This result is popularly known as Shannon's channel coding theorem.

Specializing this result for the multiple antenna channel (1.2), when **H** is known at the receiver, we have that  $l(\mathbf{x}; \mathbf{y}|\mathbf{H}) = \mathbb{E}_{\mathbf{H}} \left\{ \log \det \left( \mathbf{I}_{N_r} + \frac{P}{N_t} \mathbf{H} \mathbf{Q} \mathbf{H}^{\dagger} \right) \right\}$ , and the Shannon capacity of the multiple antenna channel is

$$C = \max_{tr(\mathbf{Q}) \le N_t} \mathbb{E}_{\mathbf{H}} \left\{ \log \det \left( \mathbf{I}_{N_r} + \frac{P}{N_t} \mathbf{H} \mathbf{Q} \mathbf{H}^{\dagger} \right) \right\},$$
(1.8)

where  $\mathbf{Q} = \mathbb{E}\{\mathbf{x}\mathbf{x}^{\dagger}\}\$  is the covariance matrix of the input signal  $\mathbf{x}$ . The optimization in (1.8) depends on whether the channel coefficient matrix  $\mathbf{H}$  is known at the transmitter (referred to as CSIT) or not (called CSIR). With CSIT, the Shannon capacity [7] is

$$C = \mathbb{E}_{\mathbf{H}} \left\{ \sum_{k=1}^{\min \{N_t, N_r\}} \log \left( \xi \sigma_k \left( \mathbf{H} \right) \right)^+ \right\},$$

where  $\xi$  is the Lagrange multiplier satisfying the power constraint

$$\sum_{k} (\xi - \sigma_k \left( \mathbf{X}_i - \mathbf{X}_j \right)^{-1})^+ = P,$$

and  $\sigma_k$  (**H**) is the *kth* eigenvalue of **HH**<sup> $\dagger$ </sup> indexed in the decreasing order.

On the other hand, with CSIR, when transmitter has no information about  $\mathbf{H}$ , the Shannon capacity [7] is

$$C = \mathbb{E}_{\mathbf{H}} \left\{ \log \det \left( \mathbf{I}_{N_r} + \left( \frac{P}{N_t} \right) \mathbf{H} \mathbf{H}^{\dagger} \right) \right\}.$$

Thus for large signal power P, with CSIT or CSIR, by using multiple antennas at both the transmitter and the receiver, the channel capacity grows linearly with  $\min \{N_t, N_r\}$ . The  $\min \{N_t, N_r\}$  factor is generally referred to as *spatial degrees of freedom*.

Next, we look at an alternate notion of capacity that is useful for non-ergodic channels for which Shannon capacity is zero.

#### **1.4 Outage Capacity**

The Shannon capacity formulation is useful for an ergodic multiple antenna fading channel, where in either each time slot or after a block of T time slots, an independent channel realization of **H** is drawn from a given distribution. T is generally referred to as the coherence time of the wireless channel. An ergodic model is valid for fast-fading case, where the fading channel coefficients change fast and the communication duration is long enough to get averaging over multiple independent blocks. Another model of interest is the non-ergodic or the slow-fading model, where the channel coefficients vary very slowly. To be specific, with the slow-fading model, it is assumed that at the start of the transmission, an independent realization of the channel matrix is drawn from any given distribution, but then is held fixed for the total communication duration. This model is well suited for low mobility wireless channels requiring short duration communication, where the coherence time is large enough compared to the total transmission time.

It is easy to see that the Shannon capacity of any non-ergodic channel is zero, because with increasing block length no averaging is available, and the error probability is lower bounded by a constant for any non-zero rate of transmission. To have a meaningful definition of capacity for

the non-ergodic channel, concept of outage capacity was introduced in [7], which is described as follows. Let B bits/sec/Hz be the desired rate of communication. Then channel outage at rate B is defined to be the event that the mutual information is less than B, and the outage probability is defined as

$$P_{\text{out}}(B) = \mathbb{P}(\mathsf{I}(\mathbf{x}; \mathbf{y}) < B)).$$

The outage capacity  $C_{\text{out}}(\epsilon)$  is defined to be the maximum rate of transmission B for which the outage probability is below a certain threshold  $\epsilon$ , that is,

$$C_{\text{out}}(\epsilon) := \max_{P_{\text{out}}(B) \le \epsilon} B.$$

The outage capacity can be interpreted as the maximum possible rate for which there exists a code whose probability of error can be made arbitrarily small for all but a set of **H**, whose total probability is less than  $\epsilon$ . Thus, in essence, outage capacity is the maximum rate which is guaranteed with success probability of at least  $(1 - \epsilon)$ .

The outage capacity formulation naturally extends to a wireless network and will be used to define a throughput metric for a wireless network called the transmission capacity in Chapter 2.

For the multiple antenna channel (1.2), with an ML decoder, the outage probability can be simplified to obtain

$$P_{\text{out}}(B) = \inf_{\mathbf{Q}, \mathbf{Q} \ge 0, tr(\mathbf{Q}) \le N_t} \mathbb{P}\left(\log \det \left(\mathbf{I}_{N_r} + \frac{P}{N_t} \mathbf{H} \mathbf{Q} \mathbf{H}^{\dagger}\right) < B\right),$$

where  $\mathbf{Q}$  is the covariance matrix of the transmitted vector  $\mathbf{x}$ .

For the most popular Rayleigh channel fading model, where each entry of the channel matrix **H** is i.i.d.  $\mathcal{CN}(0, 1)$  distributed, the distribution of the maximum mutual information expression  $\log \det(\mathbf{I} + \mathbf{HH}^{\dagger})$  is unknown. Consequently, finding the outage capacity of the multiple antenna channel has remained unsolved. The mutual information expression can be significantly simplified if instead of an ML decoder, we use a ZF decoder, where different data streams sent by the transmitter are decoupled before decoding. From [4] for (1.6), with  $N_t$  independent data streams, and assuming that each data stream is required to have rate B, and outage probability constraint  $\epsilon$ , the outage capacity of a  $N_t \times N_r$ ,  $N_r \ge N_t$  multiple antenna channel with ZF decoder is

$$C_{\text{out}}^{ZF}(\epsilon) = \max_{\mathbb{P}(\log(1+|g|^2) < B) \le \epsilon} N_t B,$$
(1.9)

where  $|g|^2$  is the signal power after zero forcing other  $N_t - 1$  signal components and hence  $|g|^2 \sim \chi^2(2(N_r - N_t + 1))$ . Thus, the outage capacity of the multiple antenna channel with ZF decoder can be found by using the CDF of a  $\chi^2$  distributed random variable with  $N_r - N_t + 1$  degrees of freedom.

After discussing the point-to-point communication scenario, we next look at the signal interactions in a wireless network, which is of primary interest in this book.

#### **1.5 Wireless Network Signal Model**

Consider a wireless network with K nodes, where the nth node's location is denoted by  $T_n$ . We assume that each node has N antennas for transmission and reception. The received signal at the